

Linhas de Transmissão - Parte 5 \$\$\$ Chapter 11



Example

11.3

Considere uma h.T. com as seguintes características:

freq. operação :  $\omega = 10^6$  rad/s

constante de atenuação :  $\alpha = 8$  dB/m

" " fase :  $\beta = 1$  rad/m

impedância característica :  $Z_0 = 60 + j40 \Omega$

comprimento :  $l = 2$  m

fonte :  $V_g = 10 \angle 0^\circ$

impedância da fonte :  $Z_g = 40 \Omega$

" carga :  $Z_L = 20 + j50 \Omega$

a) Impedância de entrada,  $Z_{in}$

b) corrente ~~no longo~~ da linha em  $z=0$ .

c) " no meio da linha.

a)  $1 \text{ Np} = 8.686 \text{ dB}$

$1 \text{ Np} \equiv 20 / \ln(10) = 8.685889638 \text{ dB}$

$1 \text{ dB} \equiv 0.115129255 \text{ Np}$

$\alpha = \frac{8 \text{ dB}}{\text{m}} \cdot \frac{0.115129255 \text{ Np}}{\text{dB}}$

$\alpha = 0.921 \text{ Np/m}$

conversão p/  $\alpha = 0 \text{ Np/m}$

$\gamma = \alpha + j\beta = 0.921 + j1 \text{ m}^{-1}$

$\gamma l = (0.921 + j1) \cdot 2 = 1.84 + j2$

$$\tanh(a+b) = \frac{\tanh(a) + \tanh(b)}{1 + \tanh(a) \cdot \tanh(b)}$$

$$= \frac{\tanh(1.84) + j \tan(2)}{1 + \tanh(1.84) \cdot j \tan(2)}$$

$\tan(1.84 + j2) = 1.0327607 - j0.0394534$  \*



Assim, a impedância  $Z_{in}$ :

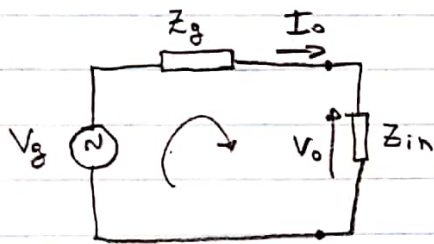
$$Z_{in} = Z_0 \cdot \frac{Z_L + Z_0 \tanh(\gamma l)}{Z_0 + Z_L \tanh(\gamma l)}$$

$$Z_{in} = (60 + j40) \cdot \left[ \frac{20 + j50 + (60 + j40) \cdot (1,0327607 - j0,0394534)}{60 + j40 + (20 + j50) \cdot (1,0327607 - j0,0394534)} \right]$$

$$Z_{in} = 60,25 + j38,79 \ \Omega \quad *$$

b) Corrente de entrada em  $z=0$

utilizando o circuito equivalente



$$V_g = I_0 Z_g + V_0$$

$$V_0 = I_0 Z_{in}$$

$$V_g = I_0 (Z_g + Z_{in})$$

$$I_0 = \frac{V_g}{Z_g + Z_{in}} = \frac{10 \angle 0^\circ}{40 + 60,25 + j38,79} = \frac{10 \angle 0^\circ}{100,25 + j38,79} = \frac{10 \angle 0^\circ}{107,493 \angle 21,153^\circ}$$

$$I_0 = 0,09303 \angle -21,153^\circ$$

$$I_0 = 93,03 \angle -21,153^\circ \text{ mA} \quad *$$

c) Encontre a corrente em qualquer ponto.

$$I_z(z) = \frac{1}{Z_0} \left[ V_{s0}^+ e^{-\gamma z} - V_{s0}^- e^{\gamma z} \right] \quad \text{da equação (2)}$$

precisamos encontrar  $V_{s0}^+$  e  $V_{s0}^-$

$$V_s(z=0) = V_0 = V_{s0}^+ + V_{s0}^-$$

$$I_s(z=0) = I_0 = \frac{V_{s0}^+}{Z_0} - \frac{V_{s0}^-}{Z_0}$$

$$V_{s0}^+ = V_0 - V_{s0}^-$$

$$V_{s0}^- = \frac{V_{s0}^+}{Z_0} - I_0 Z_0$$

$$V_{s0}^+ = V_0 - V_{s0}^- + I_0 Z_0$$

$$V_{s0}^+ = \frac{1}{2} (V_0 + I_0 Z_0)$$

$$V_{s0}^- = \frac{1}{2} (V_0 + I_0 Z_0) - I_0 Z_0$$

$$V_{s0}^- = \frac{1}{2} V_0 + \frac{1}{2} I_0 Z_0 - I_0 Z_0$$

$$V_{s0}^- = \frac{1}{2} (V_0 - I_0 Z_0)$$



Do circuito equivalente do item (b)

$$V_o = I_o Z_{in}$$

$$V_o = 93,03 \angle -21,153^\circ \cdot (60,25 + j38,79)$$

$$V_o = 93,03 \angle -21,153^\circ \cdot 71,657 \angle 32,774^\circ$$

$$V_o = 6,666 \angle 11,621^\circ$$

Sabemos também q/ :

$$V_s(z) = V_{s0}^+ e^{-\gamma z} + V_{s0}^- e^{\gamma z}$$

A tensão e corrente  $V_o$  e  $I_o$  correspondem a  $V_s(z=0)$  e  $I_s(z=0)$ , assim basta substituímos

$$V_{s0}^+ = \frac{1}{2} (V_o + Z_0 I_o) \quad \text{ver ao lado}$$

$$V_{s0}^- = \frac{1}{2} (V_o - Z_0 I_o)$$

$$V_{s0}^+ = \frac{1}{2} (6,666 \angle 11,621^\circ + (60 + j40) \cdot 93,03 \angle -21,153^\circ \text{ mA})$$

$$V_{s0}^+ = 6,687 \angle 12,08^\circ \quad *$$

$$V_{s0}^- = \frac{1}{2} (6,666 \angle 11,621^\circ - (60 + j40) \cdot 93,03 \angle -21,153^\circ \text{ mA})$$

$$V_{s0}^- = 0,0518 \angle 260^\circ \quad *$$

No meio da linha  $\Rightarrow z = l/2 \Rightarrow \gamma z = 0,921 + j1$

$$I_s(z=l/2) = \frac{V_{s0}^+}{Z_0} e^{-\gamma z} - \frac{V_{s0}^-}{Z_0} e^{\gamma z}$$



$$I_s(z=l/2) = \frac{(6,687 e^{j12,08^\circ}) \cdot e^{(-0,921 - j1)}}{60 + j40} - \frac{(0,0518 e^{j260^\circ}) e^{0,921 + j1}}{60 + j40}$$

$$I_s(z=l/2) = \frac{6,687 e^{j12,08^\circ} \cdot e^{-0,921} \cdot e^{-j57,3^\circ}}{72,1 e^{j33,69^\circ}} - \frac{0,0518 e^{j260^\circ} \cdot e^{0,921} \cdot e^{j57,3^\circ}}{72,1 e^{j33,69^\circ}}$$

$$I_s(z=l/2) = 0,0369 e^{j78,91^\circ} - 0,001805 e^{j283,61^\circ}$$

$$= 6,673 - j34,456 \text{ mA}$$

$$I_s(z=l/2) = 35,10 \angle 281^\circ \text{ mA} *$$

Exercício: Sugestão q/ os alunos resolverem

1) LT de 40m,  $V_g = 15 \angle 0^\circ$  Vrms,  $Z_0 = 30 + j60 \Omega$   
 $V_L = 5 \angle -48^\circ$  Vrms.

Supondo linha casada com o carga, calcule

# practice exercise  
11.3

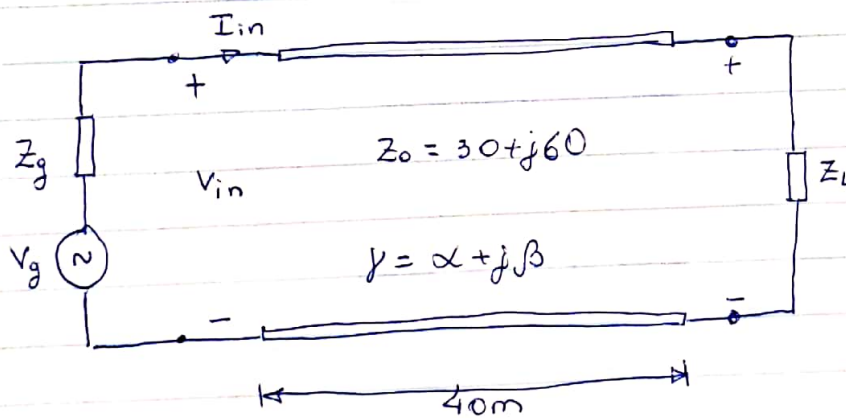
- a)  $Z_{in}$
- b) corrente de entrada  $I_{in}$  e tensão  $V_{in}$
- c) constante de prop.  $\gamma$

Resposta:

(a)  $30 + j60 \Omega$

(b)  $0,112 \angle -63,43^\circ$  A       $7,5 \angle 0^\circ$  Vrms

(c)  $0,0101 + j 0,2094$  /m



practice ② L.T. operando em  $f = 500 \text{ MHz}$

8/ alunos

exercise

$$Z_0 = 80 \Omega$$

11.1

$$\alpha = 0,04 \text{ Np/m}$$

$$\beta = 1,5 \text{ rad/m}$$

Encontre:  $R, L, G, \text{ e } C$

$$R = 3,2 \Omega/\text{m}$$

$$G = 5 \times 10^{-4} \text{ S/m}$$

$$L = 38,2 \text{ nH/m}$$

$$C = 5,97 \text{ pF/m}$$

example

11.2

③ Uma L.T sem distorção apresenta:

$$Z_0 = 60 \Omega$$

$$u = 0,6 c$$

$$c = \frac{1}{\sqrt{\mu_0 \epsilon_0}}$$

$$\alpha = 20 \text{ mNp/m}$$

Encontre:

$R, L, G, C, \text{ e } \lambda$  p/  $f = 100 \text{ MHz}$

$$RC = GL$$

$$\text{ou } G = \frac{RC}{L}$$

portanto:

$$Z_0 = \sqrt{\frac{L}{C}}$$

$$\alpha = \sqrt{RG} = \sqrt{\frac{R \cdot RC}{L}} = R \sqrt{\frac{C}{L}} = \frac{R}{Z_0}$$

$$\text{Assim, } \boxed{R = \alpha Z_0} \quad R = (20 \times 10^{-3}) \cdot (60) = 1,2 \Omega/\text{m} *$$

$$u = \frac{\omega}{\beta} = \frac{1}{\sqrt{LC}}$$

Utilizando:

$$\frac{Z_0}{u} = \sqrt{\frac{L}{C}} \cdot \sqrt{LC} = L$$

$$L = \frac{Z_0}{u} = 333 \text{ nH/m} *$$

como :

$$G = \frac{\alpha^2}{R} = \frac{400 \times 10^{-6}}{1,2} = 333 \mu S/m \quad *$$

fazendo

$$Z_0 \cdot u = \sqrt{\frac{L}{C}} \cdot \frac{1}{\sqrt{LC}}$$

$$Z_0 u = \frac{1}{C}$$

$$C = \frac{1}{Z_0 u} = \frac{1}{0,6 \cdot (3 \times 10^8) 60} = 92,59 \text{ pF/m} \quad *$$

$$\lambda = \frac{u}{f} = \frac{0,6 (3 \times 10^8)}{10^8} = 1,8 \text{ m} \quad *$$